

# Simulation Challenges of SiC MOSFET Switching Performance and Reliability

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**Abstract**—This paper describes a holistic SiC MOSFET simulation flow with a focus on switching and reliability performance in the form of Short-Circuit validation. Several key physical aspects native to SiC MOSFETs are elucidated in TCAD simulations. A novel link between TCAD and SPICE is demonstrated. This link leads into a full SPICE model calibration simultaneously to typical power converter operating regions and Short-Circuit reliability.

**Keywords**—Silicon-Carbide (SiC), power MOSFET, TCAD, SPICE, Short-Circuit

## I. INTRODUCTION

Silicon-Carbide (SiC) is now a prevalent technology in the power electronic space. The fast growing demand for electric vehicles (EV) is driving the market for high efficiency semiconductor power solutions. The electricity requirements of these EVs are also driving the demand for solar inverters and wind farms to generate electricity from renewable sources. The higher power density and faster switching native to these applications presents new design challenges for designers of SiC MOSFET power modules. Without complete simulation flows, costly fabrication and test-based design cycles are often needed.

There are many distinct challenges to simulating SiC MOSFETs from TCAD during technology development to SPICE in application simulation. The paper looks at several aspects native to SiC MOSFET modeling and simulation with a focus on the Short-Circuit test (SC). This test is a critical success criterion in 3-phase inverters for EV motor control to protect the battery and the components of the vehicle in case of loss of control of the converter. SC type I is common SC test simplification where the Device Under Test (DUT) is gated on into a faulty event with shorted load.

Methodologies are described that link device modeling and simulation between TCAD and SPICE in a

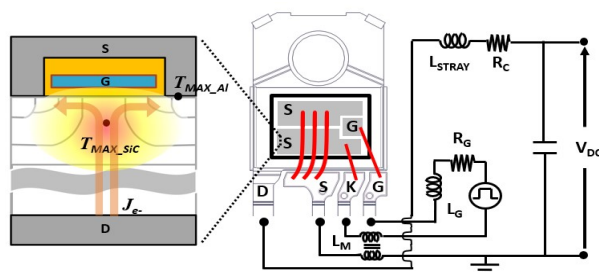


Fig. 1. Schematic description of the mixed-mode simulation for SC test in SiC MOSFETs. Schematic cross section indicates current flow and maximum temperature locations in SiC and Al regions ( $T_{MAX\_SiC}$ ,  $T_{MAX\_Al}$ )

cohesive flow. Section II details challenges in SiC MOSFET TCAD simulation. Sections III describes a technique to link TCAD to SPICE in a physical and automated manner. Section IV provides experimental results and validation of the SPICE models. Finally, section V states the conclusions and future work.

## II. NEW ELECTRO-THERMAL SiC MOSFET TCAD PHYSICS MODELS

For those TCAD users transitioning from Silicon to SiC power MOSFETs, most of the novelty encountered in physics models is directly or indirectly related to the one of the following aspects:

- traps and trapped charges at or near the SiC/Oxide interface
- non-ionized impurities and crystal defects acting as traps and recombination centers in the bulk
- anisotropy for electrical and thermal characteristics

Besides these three phenomena, all the rest of parameters in the electrical and thermal physical models are normally adapted from Silicon to 4H-SiC, which is a wide-bandgap material featuring high critical electric field, saturation velocity, and thermal conductivity.

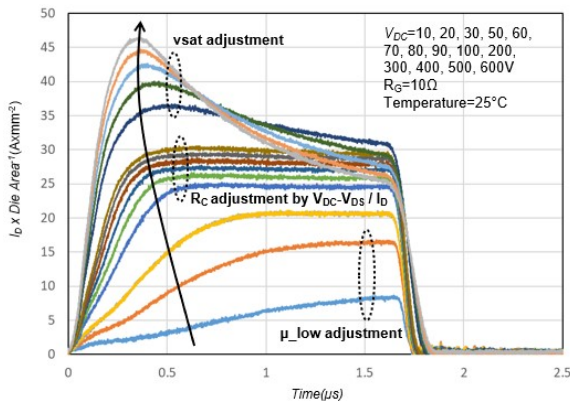


Fig. 2. Set of measured  $I_D$ - $V_{DS}$  waveforms from SC test in a wide range of  $V_{DC}$ . Different  $V_{DC}$  levels allow different adjustments in TCAD models and circuit parasitics

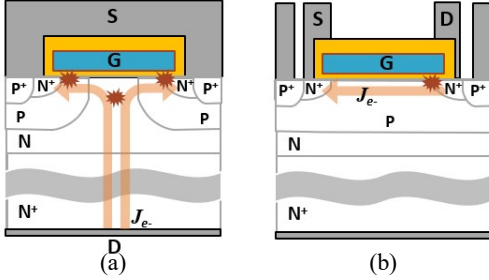


Fig. 3. Schematic description of cross section in (a) vertical SiC MOSFET and (b) lateral SiC MOSFET test structure indicating current flow and current saturation locations

Regarding the channel mobility at low electric field ( $\mu_{low}$ ), some contributions like the Coulomb scattering, are related to the existence of charge at the SiC/Oxide interface [1]. To properly calibrate the density of traps at SiC/Oxide interface, DC transfer characteristics are used. Differently, the trap parameters related to capture/emission require the support from dynamic testing. An accurate dynamic model for traps is fundamental to capture effects like BTI and  $V_{th}$  hysteresis [2].

As per the non-ionized impurities, the incomplete ionization model for dopants is of utmost importance to capture switching losses in TCAD simulations [3]. Also related to implanted dopants, the implanted regions may imply a certain level of damage in the crystal, thus generating recombination centers for free carriers. This requires an additional complexity in defining carrier lifetime models in different regions of the SiC bulk [4]. In combination of the mobility models, an accurate lifetime prediction is crucial to replicate the third quadrant conduction and reverse recovery.

Finally, the orientation effect for the hexagonal crystal structure of 4H-SiC should be accounted for in the TCAD simulations. In this sense, the anisotropic carrier heating and drift velocity lead to the anisotropy of impact ionization and breakdown voltage. Similarly, the anisotropy in phonon thermal transport, thermal resistivity, and specific heat needs special attention in scenarios with self-heating [4].

As a matter of example, the Short-Circuit test (SC) is one of the most challenging switching events to be simulated [5]. It entangles practically all the previous physical models and additional ones related to non-permanent high gate current [6]. In fact, the extreme local temperatures ( $T_J = T_{MAX\_SiC} > 1300K$ ) enhance the thermionic effect in such a way that tunneling models, recommended for very thin oxides ( $< 3nm$ ), are suitable to predict gate current in SiC MOSFETs with thick oxide ( $> 40nm$ ). Moreover, SC simulation also needs the use of the so called mixed-mode combining TCAD and SPICE elements. The latter accounts for the external circuit that provides electrical excitation as well as for all the parasitic elements from device package, PCB boards and passive components [5].

A schematic description of our mixed-mode simulations is represented in Fig. 1 without explicitly displaying the package parasitics. The parasitic elements in the gate and power loops can be extracted from other finite-element simulations or by simple analysis of measured waveforms. In the second option, a set of  $V_{DS}$  and  $I_D$  waveforms from onsemi's advanced EliteSiC MOSFET technology for a wide  $V_{DC}$  range is convenient as seen in Fig. 2. For  $V_{DC} < 30V$ , curves assist in the calibration of  $\mu_{low}$  related models, being more efficient in reducing self-heating effect than curve tracer pulses ( $> 100\mu s$ ). For  $40V < V_{DC} < 100V$ , curves are ideal to calculate the power loop series resistance  $R_C$ . For  $V_{DC} > 100V$ , the self-heating effect and current saturation are dominant effects. At any  $V_{DC}$  range, the power loop stray inductance ( $L_{STRAY}$ ) can be easily extracted by  $\Delta V_{DS} = L_{STRAY} \cdot dI_D/dt$ .

As a part of the calibration methodology, dedicated test structures are also desired. Hence, lateral SiC MOSFETs depicted in Fig. 3 help to disentangle the role of the channel and JFET regions in the current saturation with separated calibration of the carrier velocity saturation models. By using the Canali model for current saturation with high field mobility [7], the velocity saturation can be adjusted to match  $I_D$  peak in SC as shown in Fig. 4.

### III. TCAD TO SPICE LINK

It is well known that TCAD, especially mixed-mode, has long simulations time that can be prohibitive for large scale application circuits with power modules. On the other hand, SPICE provides much faster simulation time however at the expense of physical accuracy. Recently, advance physically based models for SiC MOSFETs have been developed that retain reasonable physics of the technology including process parameters [8], [9]. Such models provide a link back to the full physical accuracy of TCAD through a custom developed flow in Sentaurus Work Bench presented here called SPICE Assist as shown in Fig. 5 [10].

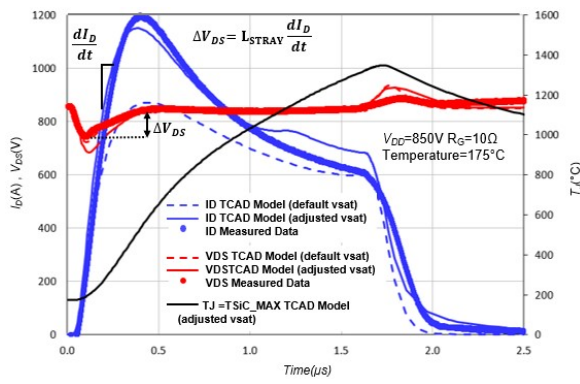


Fig. 4. Measured vs. TCAD  $I_D$ - $V_{DS}$  waveforms from SC test at the more stringent conditions

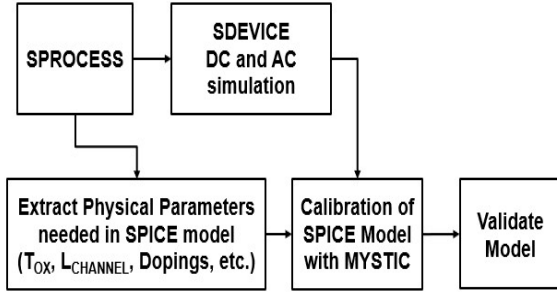


Fig. 5. SPICE Assist flow within Sentaurus workbench

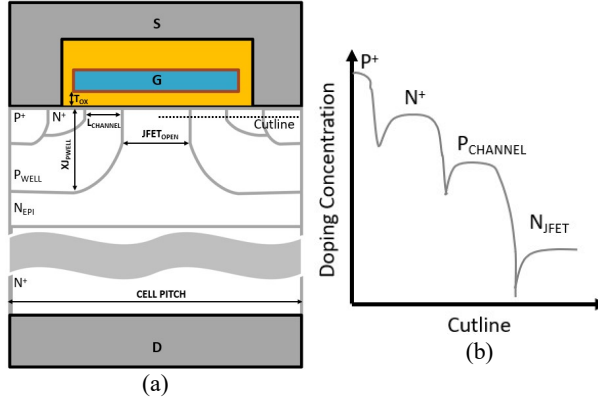


Fig. 6. SiC MOSFET process parameters used in SPICE Assist flow shown in (a) cross section and (b) doping profile

Following the SPROCESS step, the essential process parameters from the device structure are automatically extracted via a TCL script within the SPROCESS command file. The script facilitates the extraction of all necessary process parameters, such as  $T_{OX}$ ,  $L_{CHANNEL}$ ,  $P_{CHANNEL}$ ,  $N_{JFET}$ , and others as shown in the cross section and doping profile in Fig. 6. These extracted process parameters are fed into an automated MYSTIC calibration step. The MYSTIC step calibrates the model to the data generated by SDEVICE. SVISUAL then provides model validation plots as shown in Fig. 7. SPICE Assist can be used to generate models practically instantaneously that track changes in the technology through SPROCESS. This is particularly useful for application engineers to feed back application performance to technology developers early in the technology development phase.

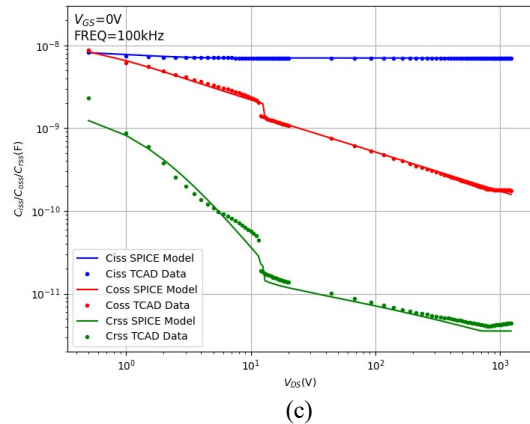
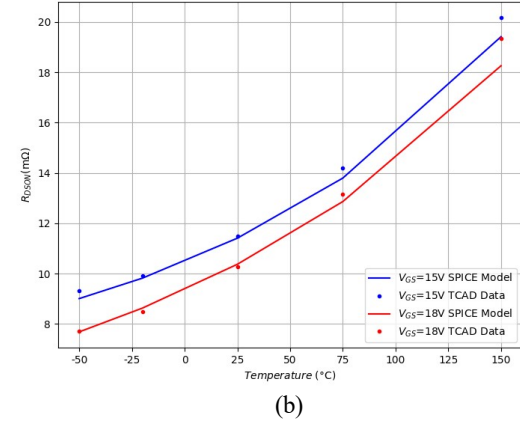
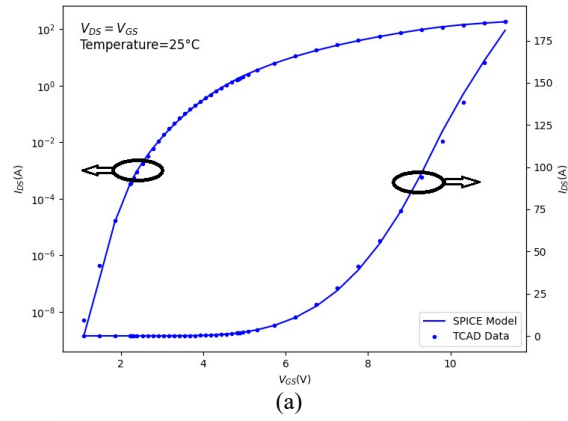


Fig. 7. SPICE Assist validation for (a)  $I_D$ - $V_{GS}$ , (b)  $R_{DS(on)}$ , and (c) Capacitance- $V_{DS}$

#### IV. SPICE MEASUREMENT VALIDATION

The final step in the overall simulation flow is to build and validate SPICE models to measurements on the same EliteSiC MOSFET from section II. Starting with the SPICE Assist models, a full calibration to measured data including the Short-Circuit test described in section II is performed. Fig. 8 displays the typical model matching for  $I_D$ - $V_{DS}$ ,  $R_{DS(on)}$ , and capacitance, where the  $I_D$ - $V_{DS}$  range is considered low power. Conventional power MOSFET model extraction focuses solely on this low power region since curve tracers are not able to measure I-V in high power regions typical of device operation in Short-Circuit. Model parameters including self-heating for the channel, JFET, and drift resistance are extracted to match the low power data. The problem

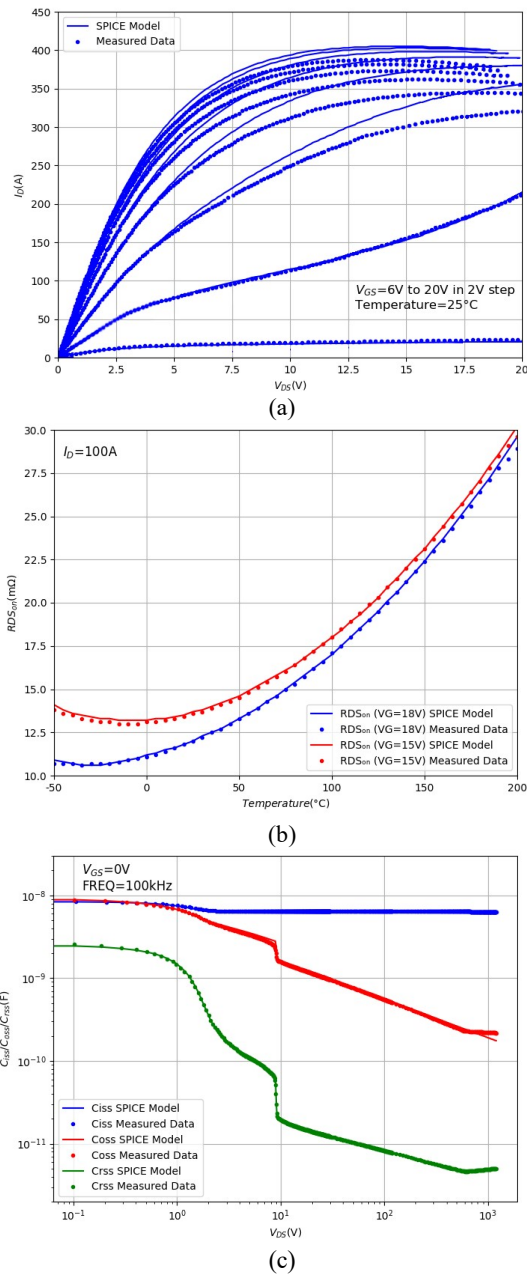


Fig. 8. Measurement validation of SPICE model for (a)  $I_D$ - $V_{DS}$ , (b)  $R_{DS(on)}$ , and (c) Capacitance- $V_{DS}$

with this approach is, the modeling engineer has no information about the very high drain current due to the simultaneously high gate and drain voltages in SC. The low power model is used somewhat blindly to simulate SC performance. If there is an imbalance between self-heating parameters, channel velocity saturation and its temperature coefficient, and temperature coefficients of the JFET, the SC drain current will not match its measurement. With the inclusion of SC data from the start of the model extraction, the right balance for the interdependence of the model parameters on low and high power regions can be found. Fig. 9 displays the SPICE model match to the SC data before using SC data and after. It should be noted accurate capacitance modeling as shown in Fig. 8(c) is also required to capture the SC switching waveforms. The peak of the

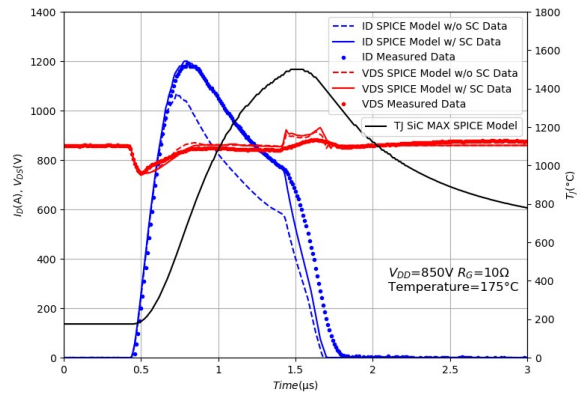


Fig. 9. Measured vs. SPICE  $I_D$ - $V_{DS}$  waveforms from SC test

SC current is predominantly affected by the channel velocity saturation and self-heating parameters. Not surprisingly, this is similar to the TCAD findings in section II. The SPICE simulated  $T_J$  matches the TCAD results reasonably well. The accurate SPICE modeling of the SC  $I_D$  together with low power  $I_D$  including  $R_{DS(on)}$ , the capacitance, and the  $T_J$  enables accurate application and reliability simulation.

## V. CONCLUSION

This paper presents a critical study of physical phenomena that affect the TCAD and SPICE simulation of onsemi's advanced SiC MOSFET technology with a focus on Short-Circuit reliability. The presented simulation flows demonstrate how TCAD, SPICE, and their linkage can be used to enable technology developers and application engineers to work together to produce advanced reliable state of the art SiC MOSFET technologies.

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