

# Relationship between capacitance and conductance in MOS capacitors

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**Abstract**—In this work, we describe how the frequency dependence of conductance (G) and capacitance (C) of a generic MOS capacitor results in peaks of the functions  $G/\omega$  and  $-\omega dC/d\omega$ . By means of TCAD simulations, we show that  $G/\omega$  and  $-\omega dC/d\omega$  peak at the same value and at the same frequency for every bias point from accumulation to inversion. We illustrate how the properties of the peaks change with the semiconductor doping ( $N_D$ ), oxide capacitance ( $C_{OX}$ ), minority carrier lifetime ( $\tau_g$ ), interface defect parameters ( $N_{IT}$ ,  $\sigma$ ) and majority carrier dielectric relaxation time ( $\tau_r$ ). Finally, we demonstrate how these insights on  $G/\omega$  and  $-\omega dC/d\omega$  can be used to extract  $C_{OX}$ ,  $N_D$  and  $\tau_g$  from InGaAs MOSCAP measurements

**Keywords**—Characterization, extraction technique, MOS, multi-frequency, C-V, G-V, minority carrier lifetime, oxide capacitance, doping

## I. INTRODUCTION

Measuring and analyzing the impedance of the metal-oxide-semiconductor (MOS) system has played a central role in the development of MOS structures in electronic, photovoltaic and photocatalytic devices. The impedance/admittance modulus ( $Z$ ,  $Y$ ) and phase angle ( $\theta$ ) of the MOS capacitor are typically measured over the frequency range from 20 Hz to 1 MHz, and resolved into capacitive (C) and conductive (G) elements. The behavior of C and G as a function of gate voltage, temperature and frequency are used to investigate and quantify physical parameters and defects of the MOS system [1].

It was recently shown using experimental results, simulations and mathematical analysis, that for any MOS capacitor in inversion, functions of the capacitance ( $-\omega dC/d\omega$ ) and conductance ( $G/\omega$ ) are related at all angular frequencies ( $\omega=2\pi f$ ) and that the two functions both exhibit a maximum value at the transition frequency ( $f_m$ ). In addition, the values of  $G/\omega$  and  $-\omega dC/d\omega$  are equal at  $f_m$  [2].

The objective of this work is to extend on [2], to show that the functions  $G/\omega$  and  $-\omega dC/d\omega$  are related in all bias regions of the MOS C-V response and that this holds for ideal MOS systems and for non-ideal MOS structures with interface traps.

## II. THE $G/\omega$ AND $-\omega dC/d\omega$ RELATIONSHIP: VALIDATION

To demonstrate the relationship between the C and G parameters, we consider the case of an InGaAs MOS capacitor (53% In). It is important to emphasize that the relationship is expected to hold for all MOS structures. InGaAs was selected as an example MOS system for the following reasons: (a) based on the energy gap ( $E_g=0.74$  eV) and typical  $\tau_g$  values, InGaAs MOS capacitors can exhibit an inversion response at room temperature [3], (b) MOS systems based on wider band gap semiconductors, (e.g., Si, GaN), and typical values  $\tau_g$ , exhibit peak frequency values ( $f_m$ ) in inversion which are typically below 1 mHz, making difficult the measurements, (c) the C and G functions also have coincident peaks in the

GHz regime, of relevance to RF and mm wave applications, where In(Ga)As devices are often employed [4].

Figure 1(a) and 1(b) report the multi-frequency C-V and G-V responses (1 kHz to 1 MHz) of an  $In_{0.53}Ga_{0.47}As$  MOS capacitor simulated with Sentaurus device simulator [5], including Fermi-Dirac statistics, multi-valley band structure with non-parabolic corrections and Shockley-Read-Hall (SRH) generation/recombination. Figure 2 plots  $G/\omega$  and  $-\omega dC/d\omega$  in inversion ( $V_G=-3V$ ) from  $10^2$  to  $10^{14}$  Hz. The plot confirms that the two functions feature the same value at  $f=2285$  Hz, as described in [2], where the peak frequency is determined by the SRH time  $\tau_g$ . The plot also shows that the two functions exhibit a second peak value at  $f=8.53$  THz. Figure 3 plots the magnitude of the peak values of  $G/\omega$  and  $-\omega dC/d\omega$  as the gate bias is varied. The plot demonstrates how the relationship holds at all bias points in the ideal CV/GV response.

Figure 4 shows how the two peak frequencies vary with the gate voltage. The high frequency peak is always present and relates to the majority carrier dielectric relaxation time, which is discussed later. The peak due to the supply of minority carriers to the inversion layer goes to zero, as expected, when the MOS structure moves from strong inversion to depletion. This is a general relationship and holds for any MOS system.

We next consider if the  $G/\omega$  and  $-\omega dC/d\omega$  relationships still hold for the non-ideal case, namely in the presence of interface states. Figure 5 illustrates an example for a Gaussian  $D_{IT}$  profile, described by the equation  $D_{IT}(E) = N_{IT} \exp\left(-\frac{(E - E_{IT})^2}{2S_{IT}^2}\right)$ . The traps are donor type, as proposed in previous publications [e.g., [6]]. This profile is introduced into the ideal InGaAs MOS structure considered in Figs. 1. The  $D_{IT}$  introduces a frequency dependent distortion into the CV and GV response, as reported in other publications [e.g., [6]]. Plotting the  $G/\omega$  and  $-\omega dC/d\omega$  peak values versus gate voltage (Figure 6(a)) and the frequency of the peak value versus bias (Figure 6(b)), indicates that the relationship still holds for interface states. The high frequency peaks ( $> 10^{12}$  Hz) are not influenced by interface states, as expected, and are removed for clarity.

The physical meaning of the peak values of  $G/\omega$  and  $-\omega dC/d\omega$  is illustrated in Figures. 7 to 10. Figure 7 shows that the peak frequency in inversion is directly related to  $1/\tau_g$ . At high values of the minority carrier lifetime,  $f_m$  eventually saturates, when the inversion layer carrier supply rate from minority carrier density in quasi neutral region ( $n_i^2/N_D$ ) exceeds the SRH generation rate. In depletion region, where interface traps dominates the response, the peak value of  $G/\omega$  linearly depends on  $N_{IT}$  (the peak value of  $D_{IT}$ ) for  $G/\omega < C_{OX}/2$  (see Figure 8), and this is a well known result from the conductance method approach [1][7]. The peak frequency shows a linear dependence on the capture cross section (Figure 9). In accumulation, the high frequency peak is set by the majority carrier dielectric relaxation time (Figure 10)

$\tau_r = \epsilon_0 \cdot \epsilon_r \cdot \rho$ , with  $\epsilon_0$ ,  $\epsilon_r$  and  $\rho$  being the free space permittivity and the semiconductor relative permittivity and resistivity [8].

The measured/simulated admittance is resolved in the equivalent circuit shown in Figure 11(c). This circuit representation in addition to the equivalent circuit topology of the device under test (Fig 11(b)) set the relationship between  $G/\omega$  and  $-\omega C/d\omega$ . By first writing  $C$  and  $G$  in terms of  $C_{OX}$ ,  $G_S$ ,  $C_S$ , and then by calculating the corresponding maximum/minimum, it is also possible to obtain analytic expressions for the peak value and position of the  $G/\omega$  and  $-\omega C/d\omega$  functions:

$$f_m = \frac{G_S}{2\pi(C_S + C_{OX})} \quad (1.a)$$

$$\left(\frac{G}{\omega}\right)_{max} = \left(-\omega \frac{dC}{d\omega}\right)_{max} = \frac{C_{OX}^2}{2(C_S + C_{OX})} \quad (1.b)$$

where  $G_S$  and  $C_S$  are the conductance and capacitance of the semiconductor respectively. It is worth noting that at every bias point the semiconductor impedance can be modeled using a capacitor and a conductance, also in the presence of interface traps. The only assumption made in the topology reported in Figure 11(b) is that the oxide is free of traps. In fact, in the presence of border traps an R-C distributed network should be added in the circuit, as reported in [9].

### III. THE $G/\omega$ AND $-\omega dC/d\omega$ RELATIONSHIP: APPLICATION

By performing a direct comparison between  $G/\omega$  and  $-\omega C/d\omega$  from simulations and experimental results, MOS parameters can be extracted. An example is shown in Figure 11(a) based on experimental multi-frequency C-V and G-V measured from 1 kHz to 1 MHz for a Ni/Al<sub>2</sub>O<sub>3</sub>/In<sub>0.53</sub>Ga<sub>0.47</sub>As/InP structure [10]. The only region that can be analyzed is inversion, where the frequency peak of  $G/\omega$  and  $-\omega C/d\omega$  is within the measured frequency range (Figure 4). For non-silicon based MOS systems, defects in the oxide (border traps) can have a significant impact on the MOS/MOSFET characteristics [11][12]. However, in strong inversion the behavior of  $C$  and  $G$  is mainly given by the minority carriers response, so it is reasonable to neglect the effect of the border traps and to use the topology shown in Figure 11(b) also in this case.

Using  $C_{OX}=0.098 \text{ F/m}^2$ ,  $\tau_g=80 \text{ ps}$  and  $N_D=4.6 \cdot 10^{17} \text{ cm}^{-3}$  we obtain an excellent agreement between simulations and experiments (Fig 11(a)). The extracted parameters are consistent with other extraction techniques [10]. In particular,  $N_D$  agrees with the mean value of the doping concentration extracted by ECV, while  $C_{OX}$  is compatible with the oxide thickness measured by TEM images (6 nm) and assuming a Al<sub>2</sub>O<sub>3</sub> dielectric constant of 6.6, which is consistent with other experiments [13].

### IV. CONCLUSIONS

Simulations have shown that the  $G/\omega$  and  $-\omega C/d\omega$  versus frequency plots exhibit marked peaks. The value and the frequency of the peaks can be interpreted with the circuit elements of Figure 11(b) and thus eventually in terms of  $C_{OX}$ ,  $N_D$ ,  $\tau_g$ ,  $\tau_r$ ,  $D_{IT}$  and  $\sigma$ . In strong inversion, where the effect of border traps can be neglected for this analysis, it is possible to do a direct comparison of the  $G/\omega$  and  $-\omega C/d\omega$  functions

between experimental and simulated results (without traps). The triplet ( $C_{OX}$ ,  $N_D$ ,  $\tau_g$ ) that minimizes the error between these functions identifies the extracted quantities. This technique is completely general and applicable to any MOS capacitor system, the information that can be extracted from the experiments depends however on the frequency range that can be practically explored.

### ACKNOWLEDGMENT

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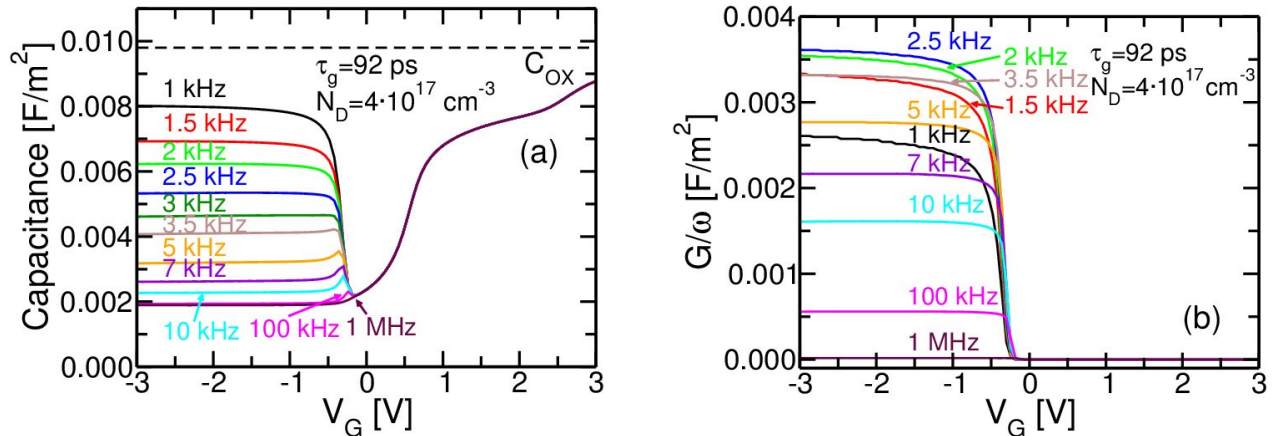


Figure 1. Simulated multi-frequency C-V (a) and  $G/\omega$  -V (b) at 300 K w/o including traps. Simulations use  $N_D=4 \cdot 10^{17} \text{ cm}^{-3}$ ,  $C_{ox}=0.0098 \text{ F/m}^2$  and  $\tau_g=92 \text{ ps}$ .

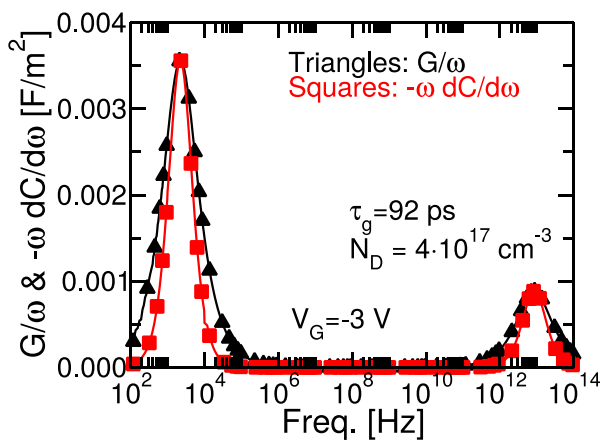


Figure 2. Simulated curves of  $G/\omega$  (triangles) and  $-\omega \cdot dC/d\omega$  (squares) extracted from the MOS structure in Figure 1 and plotted as a function of frequency in inversion ( $V_G=-3 \text{ V}$ ).

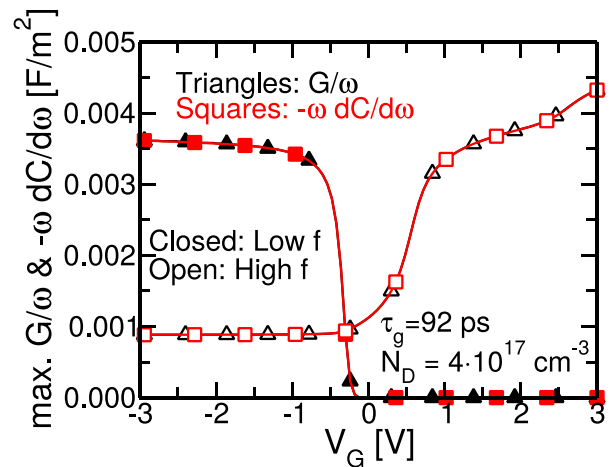


Figure 3. Peak value of  $G/\omega$  (triangles) and  $-\omega \cdot dC/d\omega$  (squares) as a function of  $V_G$  for the simulations in Figure 1

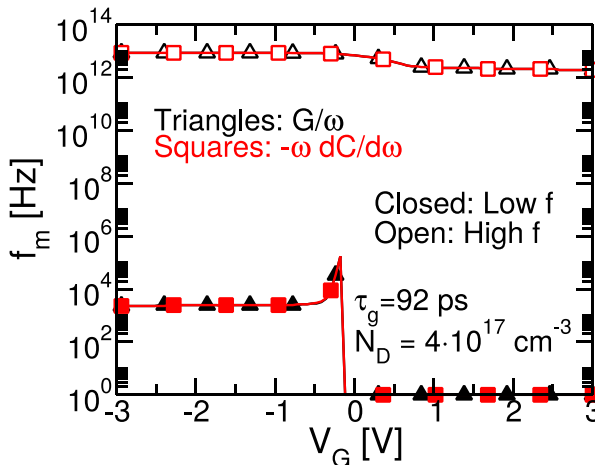


Figure 4. Frequency position of the peak value of  $G/\omega$  (triangles) and  $-\omega \cdot dC/d\omega$  (squares) as a function of  $V_G$  for the simulations in Figure 1.

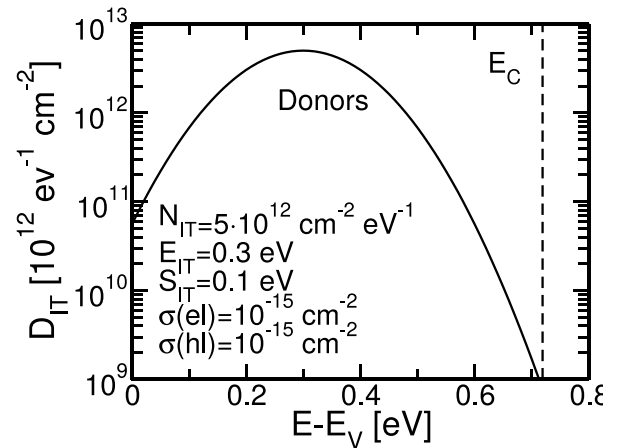


Figure 5.  $D_{IT}$  profile of donor traps included in the simulations of Figure 6. The energy is referred to the valence band of InGaAs. The figure reports also the parameter used for the Gaussian distribution and the values of the capture cross section for electron and holes.

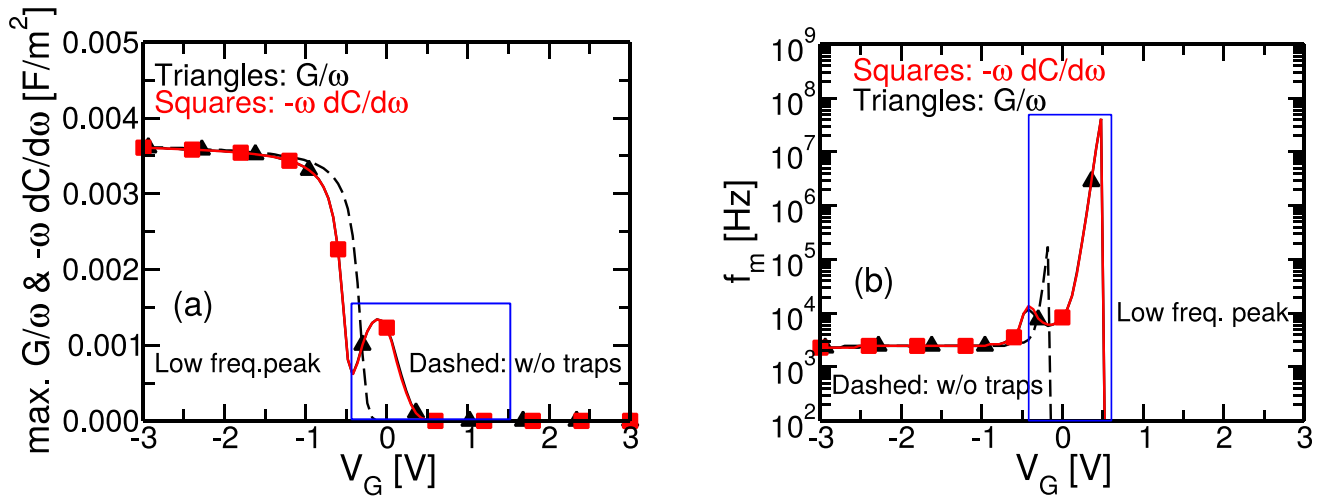


Figure 6. Simulations of the MOS structure in Figs. 1 including the Gaussian  $D_{IT}$  profile shown in Figure 5. The peak value of  $G/\omega$  and  $-\omega dC/d\omega$  versus  $V_G$  is shown in (a) and its frequency position in (b). The dashed lines are the simulation results without traps and the blue box highlights the region where the  $D_{IT}$  affects the properties of the peaks. The high frequency peaks ( $> 10^{12}$  Hz) are not impacted by  $D_{IT}$ , as expected, and are removed for clarity.

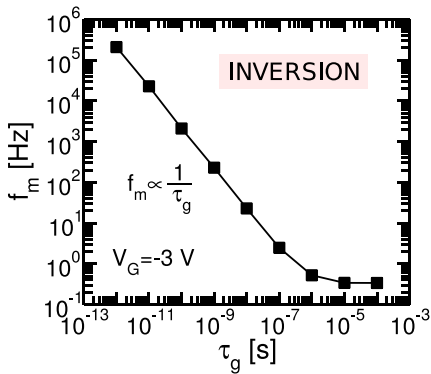


Figure 7. Simulated  $f_m$  as a function of the minority carrier life time extracted from the MOS structure in Figure 1.

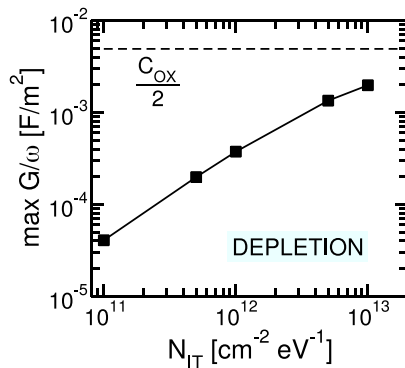


Figure 8. Simulated maximum value of  $G/\omega$  as a function of  $N_{IT}$  for the MOS structure in Figure 6. All the other parameters of  $D_{IT}$  are the ones reported in Figure 5. For high values of  $N_{IT}$ , the function tends to  $C_{ox}/2$ , which is shown with the dashed line.

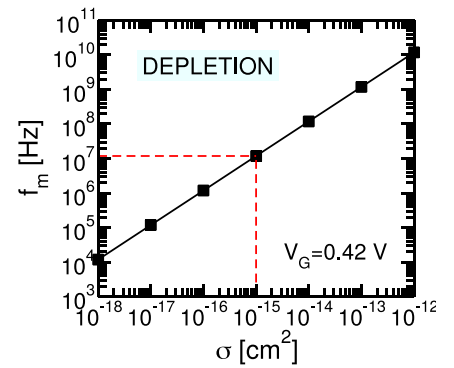


Figure 9. Simulated  $f_m$  associated to traps response of the  $D_{IT}$  in Figure 5 as a function of the capture cross section. Note that for  $\sigma = 1 \cdot 10^{-15}$   $\text{cm}^2$ ,  $f_m$  is  $\sim 10$  MHz, which is outside the typical measurement range (up to 1 MHz).

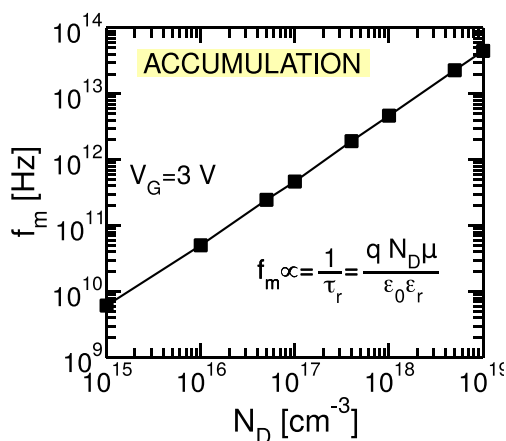


Figure 10. Simulated  $f_m$  of the majority carriers as a function of  $N_D$  from the MOS structure in Figure 1 (relevant for *mm* wave devices).

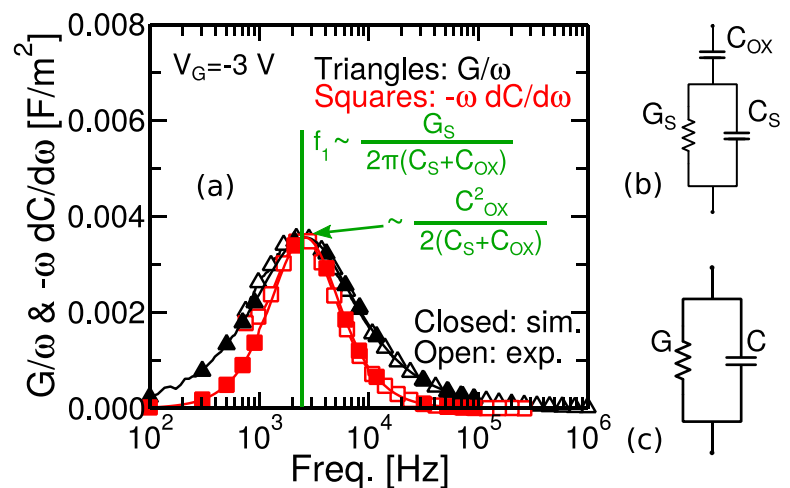


Figure 11. (a) Simulated (closed) and experimental (open) curves at  $V_G = -3$  V of  $G/\omega$  (triangles) and  $-\omega dC/d\omega$  (squares) plotted as a function of frequency. (b) Equivalent electrical circuit of a MOS structure interpreted using a parallel G-C circuit. (c) Equivalent electrical circuit of a MOS structure interpreted using a parallel G-C circuit.